
Introducing Spectrally Efficient Transmission over DWDM Optical Networks: Impact on Subsystems

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Outline

- Introduction: Modulation Formats
- Performance of Spectrally Efficient Modulation Formats
- Transceiver Design for DQPSK
- Upgrading DWDM Networks with Advanced Modulation Formats
- Performance of Electronic Equalizers for Different Modulation Formats
- Combining FEC and Modulation Formats

The Demand in Optical Networks

- Increase Capacity & Robustness
- Reduce Cost
- Optical communications: Development stages:
 - Single mode fibre : low loss @1550nm, low dispersion @1310nm
 - Optical amplifiers @1550nm: regenerate in optical domain
 - High-speed data rates: 10Gb/s → 40Gb/s → 160Gb/s
 - DWDM (Dense Wave Division Multiplex): more fibre bands (more channels)
 - Dense channel spacing, spectral efficiency
 - Advanced **signal processing**:
 - Robust, efficient → cost saving!
 - **modulation, equalization, channel coding**

Spectrally Efficient Transmission: Impact on Subsystems

- Use alternative Modulation formats
- Impact of these new modulation formats on:
 - Robustness towards channel impairments
 - Upgrading existing DWDM systems
 - Equalizer Design
 - FEC performance

Modulation Formats: Some Options

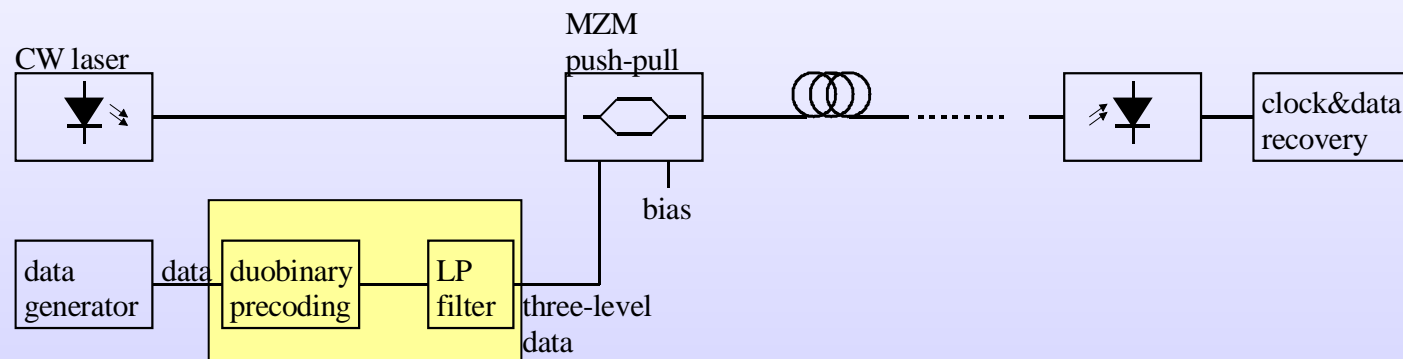
- Amplitude
 - RZ/NRZ (2-ASK with different pulse shape)
 - VSB/SSB
 - Multilevel (4-ASK, 8-ASK)
- Phase
 - DPSK
 - DQPSK
 - Multilevel PSK
- Frequency
 - MSK/FSK = minimum shift keying/ frequency shift keying
- Hybrid
 - CS-RZ, chirped RZ/NRZ, partial response (e.g. duobinary)
 - ASK-DPSK
 - QAM
 - Multiring-QPSK

Innovative Formats: Duobinary Signaling

- ✓ Reduced bandwidth (approx. Factor of 2 against NRZ)
- ✓ Highly dispersion tolerant → increase uncompensated transmission reach
- ✓ Increased spectral efficiency
- ✓ Carrier suppressed format: relaxed SBS suppression requirements
- ✓ Low implementation complexity → upgrade of conventional IM/DD possible
- Reduced back-to-back performance

Duobinary Signaling - Implementation

- Duobinary precoder = differential encoder \rightarrow IC-solution
- Duobinary filter: LP with cut-off frequency of $f_b/4$



Duobinary Encoding

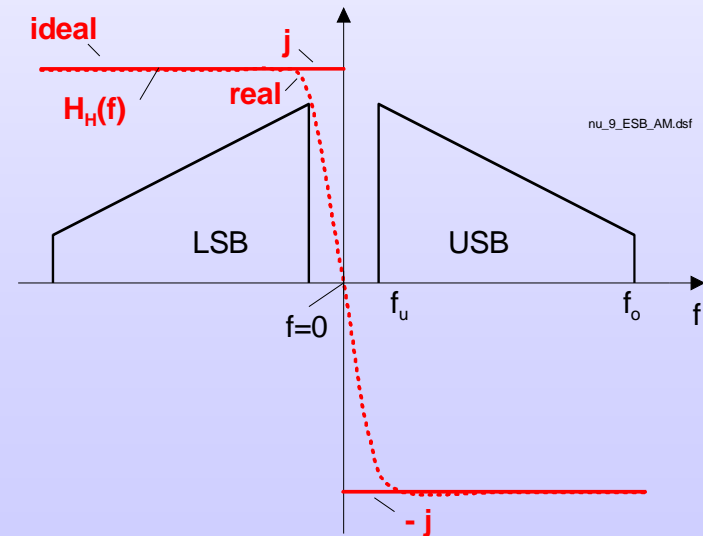
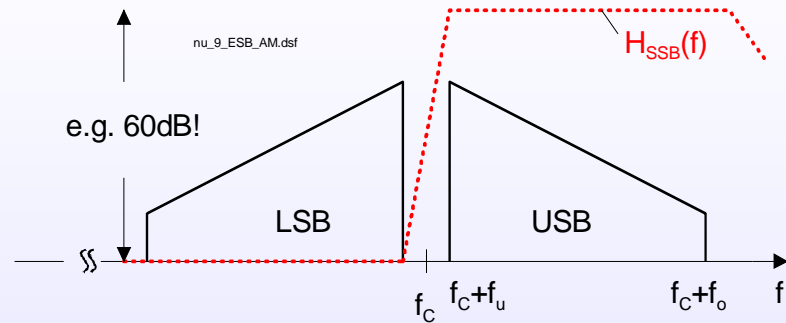
Innovative Formats: SSB/VSB

- ✓ Single Side Band (SSB) or Vestigial Side Band (VSB)
- ✓ Reduced bandwidth (approx. factor 2) → Highly dispersion tolerant
→ Increased spectral efficiency
- Medium to high implementation complexity
- For direct detection: carrier required → Extinction ratio problems! → reduced receiver sensitivity

SSB/VSB – Sideband Suppression

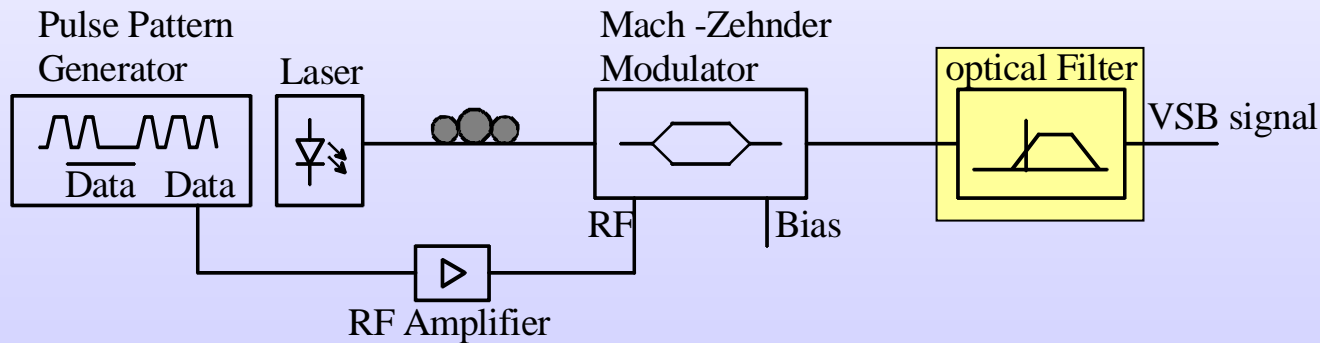
- Optical SSB/VSB filter
- Filtering in **optical** (band-pass) domain

- Hilbert Transformer = Phase response Filter
- Filtering in **electrical** (base-band) domain



SSB/VSB – Implementation: Optical Filter

- Optical filtering of conventional IM (DSB) signal
- Steep filter slope required



Binary Differential PSK (D-2-PSK)


- ✓ Symbols: +1, -1, NRZ or RZ pulse shape possible
 - ✓ RZ: → advantage in XPM environment
- ✓ Improved noise performance vs. conventional IM/DD , if balanced receiver used (3dB)
- ✓ Differential scheme → avoids coherent detection
- Differential scheme → noise penalty: 0.2dB (binary), 2dB (quaternary)
- Additional DPSK receiver (MZDI) necessary & Differential precoder necessary!

Binary DPSK

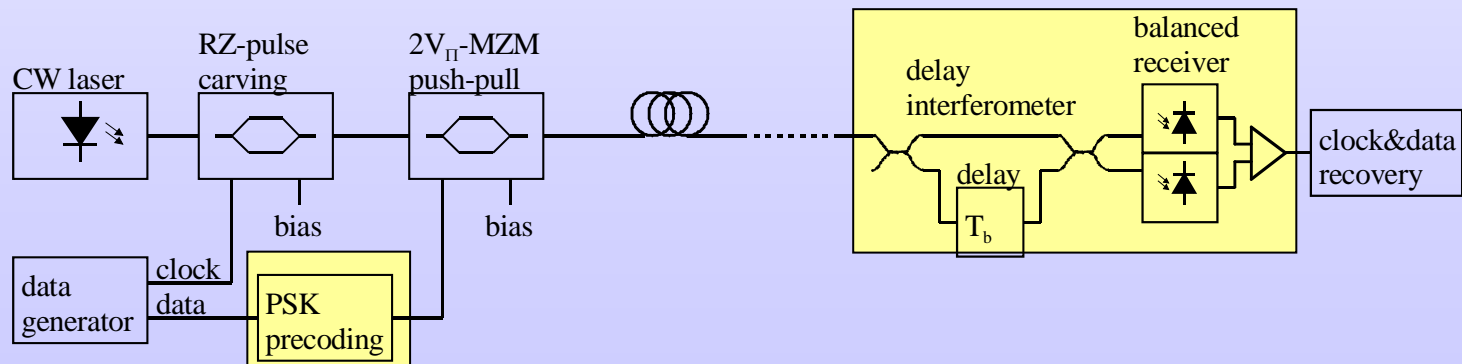
- DPSK precoder = differential encoder (same as for duobinary)
- Functionality of MZDI receiver. Phase $\phi(k) \in \{0, \pi\}$

$$\left[\cos(\omega_c t + \phi(k)) + \cos(\omega_c t - \alpha + \phi(k-1)) \right]^2 = \dots + 2 \cos(\dots) \cdot \cos(\dots) + \dots$$

$$\underbrace{\cos[\phi(k) - \phi(k-1) - \alpha]}_{=0}$$

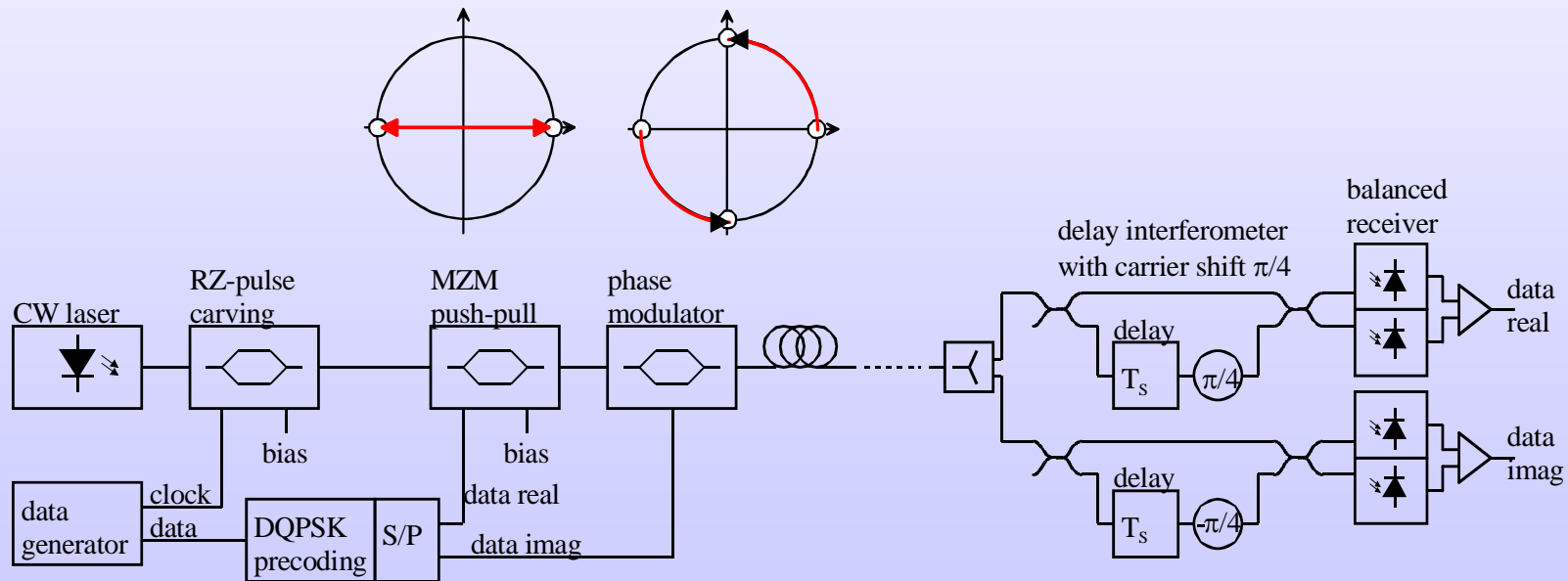
"High" if $\phi(k) = \phi(k-1)$, 

"Low" if $\phi(k) \neq \phi(k-1)$



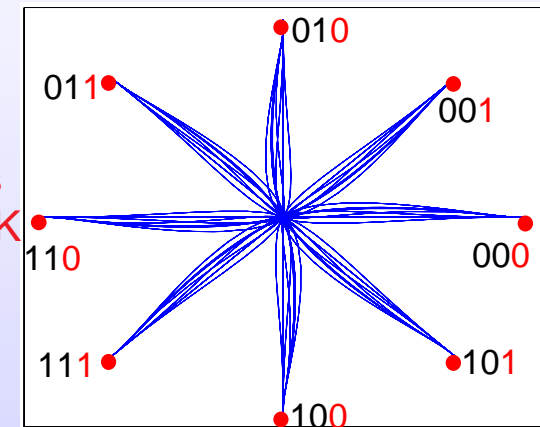
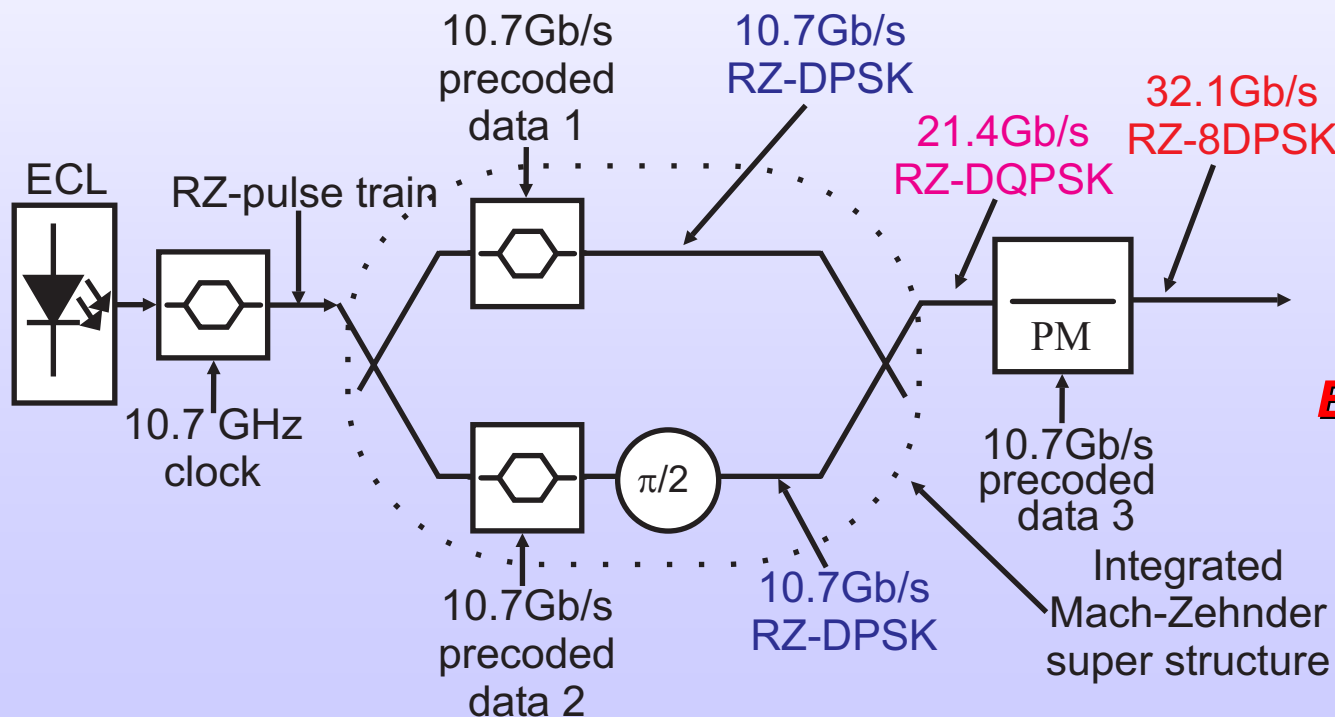
Quaternary DPSK (DQPSK)

- Our solution: serial DQPSK-transmitter → reduced complexity, relaxed phase control
- 2 parallel data streams (real & imaginary) with speed and bandwidth requirements of single data stream!
- Quaternary differential precoder required



Eight level RZ-DPSK (RZ-8DPSK)

- Upgrade from parallel RZ-DQPSK to RZ-8DPSK by cascaded phase modulator
- Driving amplitude of the phase modulator corresponds to $\frac{1}{4}V\pi$
- No additional control loop required !
- Considering 7% FEC overhead



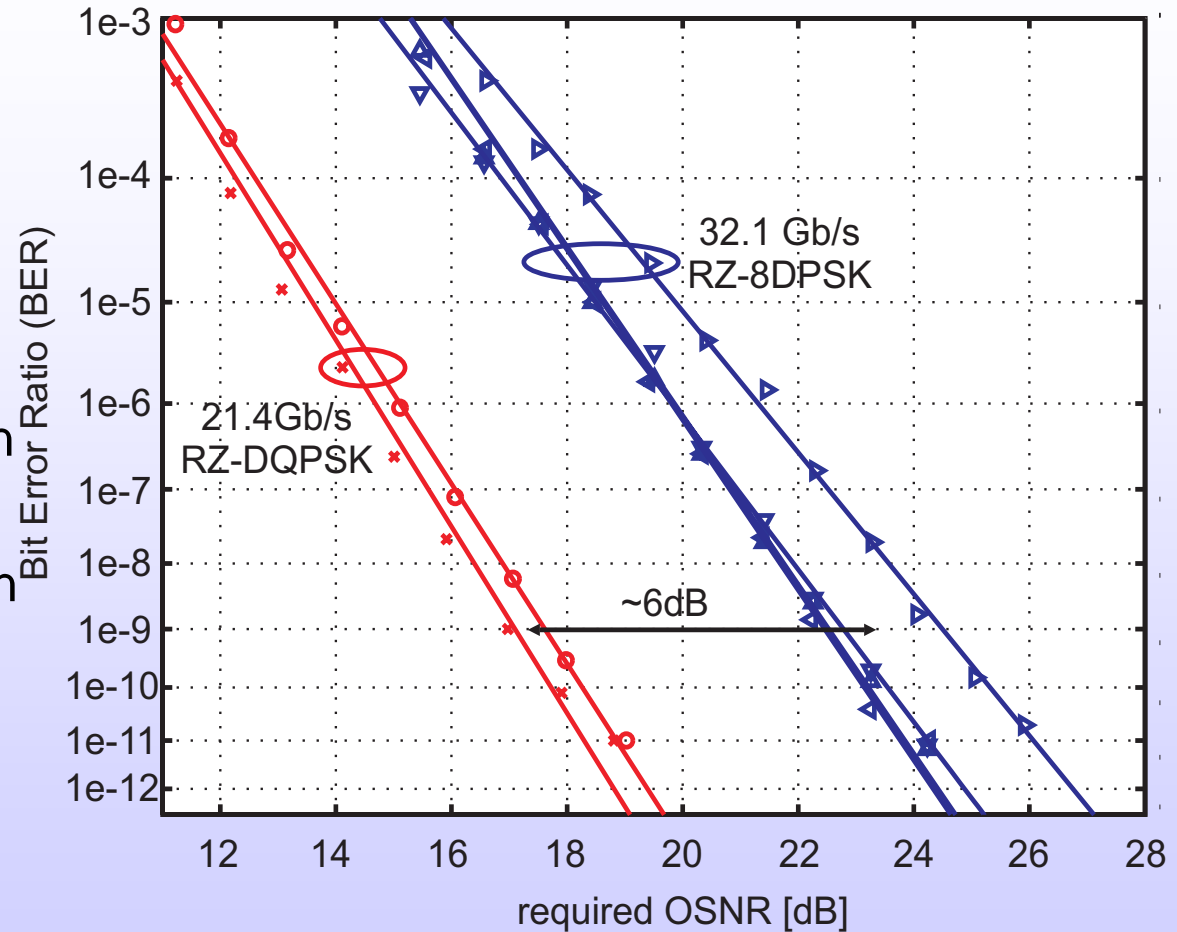
**Eight possible symbols
i.e. 3 bits per symbol
→ Data-rate increase
by 50% at same
spectral width!**

8-DPSK Experimental Results: OSNR-requirements

OSNR Requirements at $1e-9$:

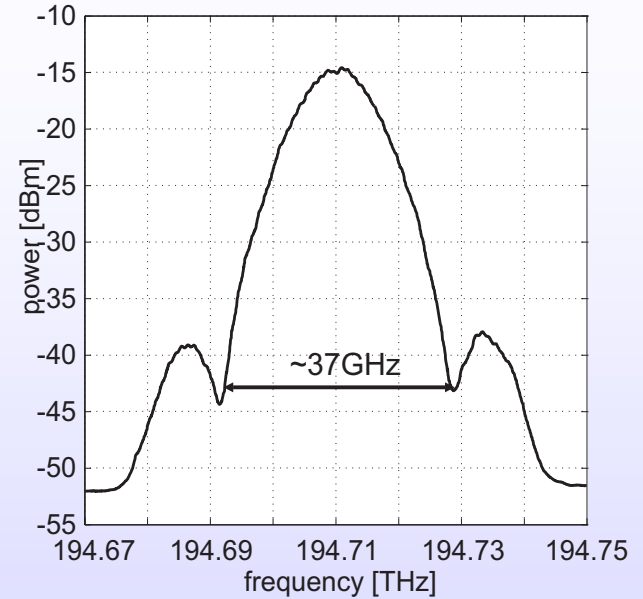
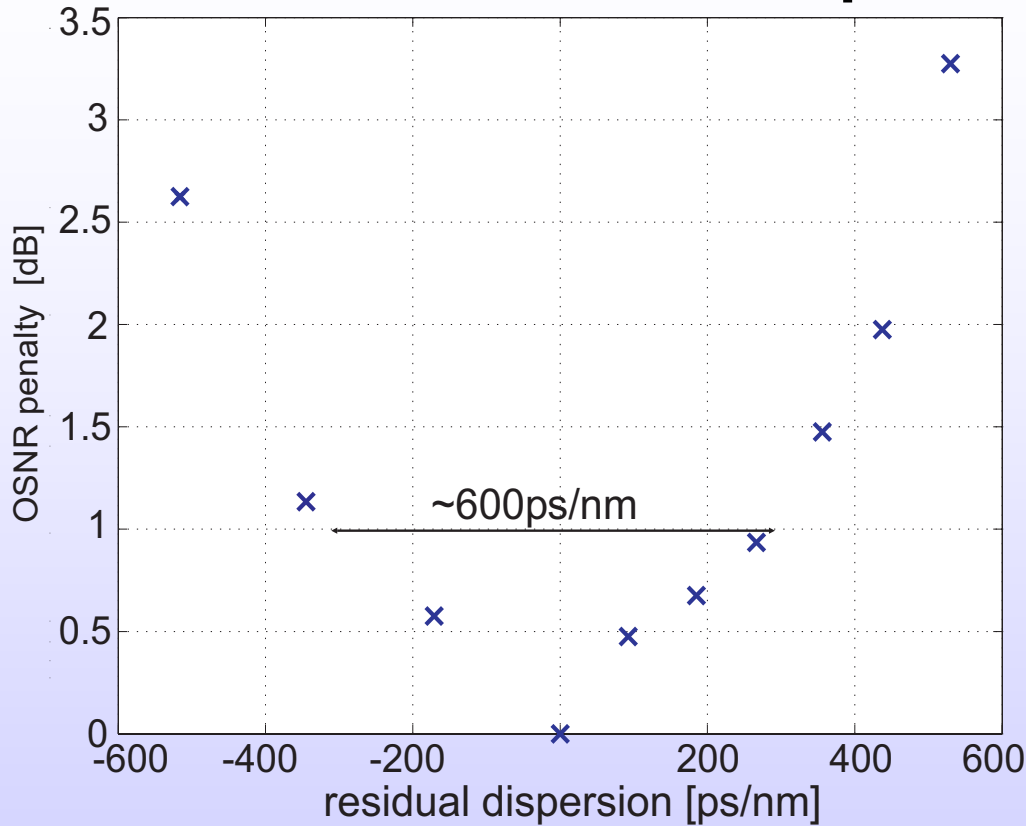
- RZ-DQPSK: 17.7dB
- RZ-8DPSK: 24.4 dB

→ At most **6.7dB penalty** due to narrower distance between adjacent symbols in the complex plane in combination with incoherent detection.



- Difference between tributaries due to imperfect receiver implementation

Measured dispersion tolerance

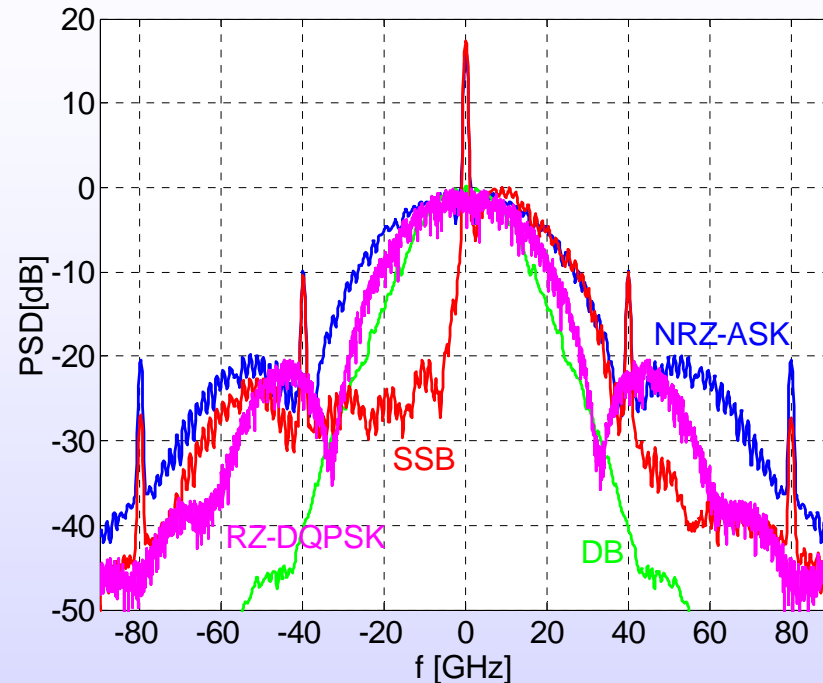


Narrow spectral width
of 37GHz for 32.1Gb/s

- Measured dispersion tolerance of 600ps/nm at 32.1 Gb/s.
- Equivalent to 35km SSMF

Modulation Formats in WDM Transmission

- Power Spectral Density of various modulation formats at 40Gb/s
- DB,SSB,RZ-DQPSK suitable for 0.8b/s/HZ i.e. 50GHZ channel spacing

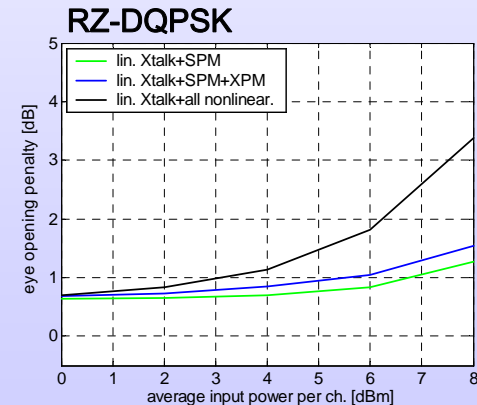
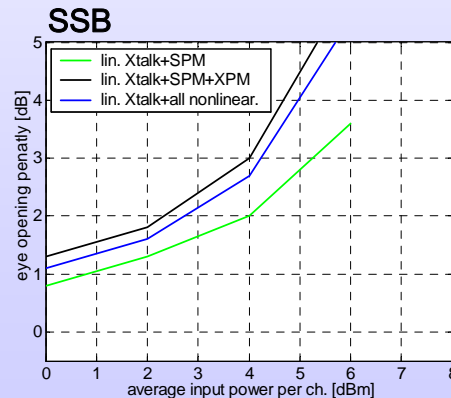
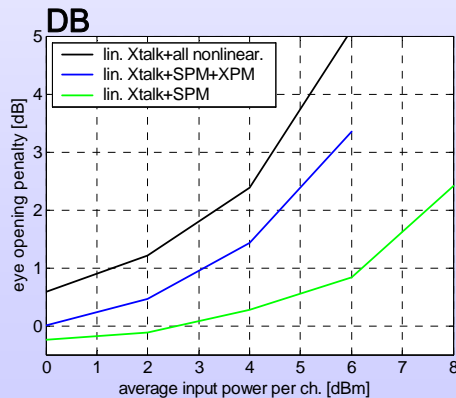


| | SSB | DB | RZ-DQPSK | NRZ-ASK |
|------------------|-----|----|----------|---------|
| Δf [GHz] | 40 | 48 | 56 | 70 |

Two-sided spectral width for a decay of 20dB at 40Gb/s data rate

Narrowband Modulation Formats in WDM: Intermodulation

| | Linear cross-talk | SPM | XPM | FWM |
|----------|-------------------|-----|-----|-----|
| DB | 😊 | | 😞 | |
| SSB | | 😞 | | 😊 |
| RZ-DQPSK | | 😊 | 😊 | 😞 |



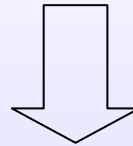
Sensitivity to nonlinear fibre effects, SSMF, 40Gb/s, $\Delta f=50\text{GHz}$, 0.8b/s/Hz

Conclusions: Performance vs. Complexity

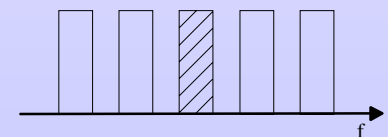
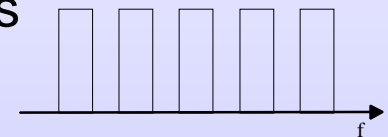
- Duobinary: low complexity, high benefit through spectral narrowing → promising candidate for system upgrade now!
- SSB/VSB: solution with optical filtering → moderate system improvement for reasonable additional complexity
- DPSK: medium complexity → high benefit through improved receiver sensitivity and WDM intermodulation robustness
- DQPSK: medium to high complexity → but doubles channel capacity
- Other advanced formats: (MSK, CPM, M-PSK, M-QAM) worth investigating in optical environment! → constraints: complexity and chirp!

Upgrading DWDM Networks with advanced Modulation Formats

Upgrade Scenario: Individual channels(groups) with advanced modulation formats coexist with “old” NRZ-ASK channels



- WDM Network with uniformly modulated channels RZ-DPSK, RZ-DQPSK, NRZ-ASK (uniform system)
- WDM Network with NRZ-ASK and individual channels upgraded with RZ-DPSK and RZ-DQPSK (mixed system!)



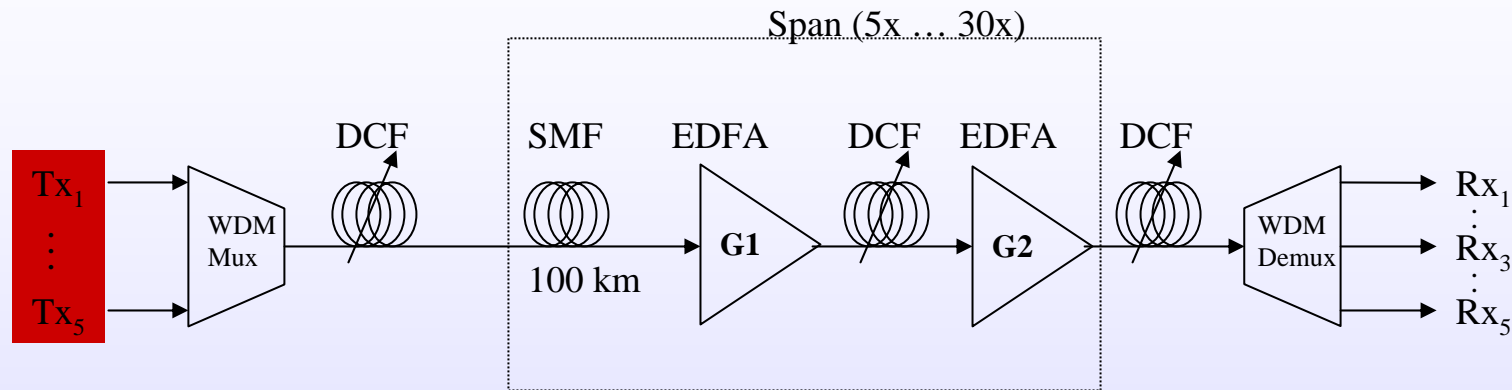
Investigations

Robustness of Modulation format

→ Dispersion tolerance

- Variation of Parameters:
 - Power
 - Transmission Distance
 - Pre compensation
 - In-line compensation
 - Post compensation
- Compute Eye Opening Penalty (EOP)
- Dispersion tolerance as a function of pre- and inline-compensation

Simulation set-up

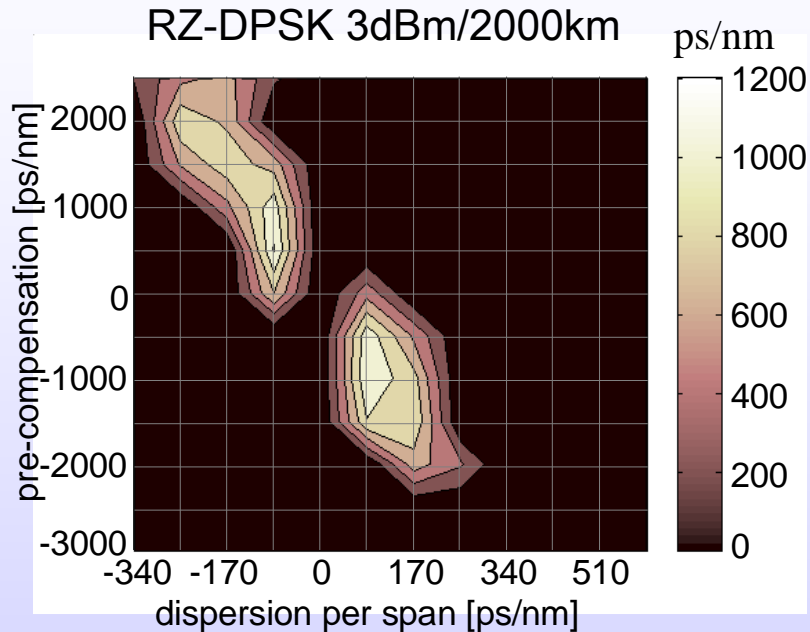


Mixed WDM-System:

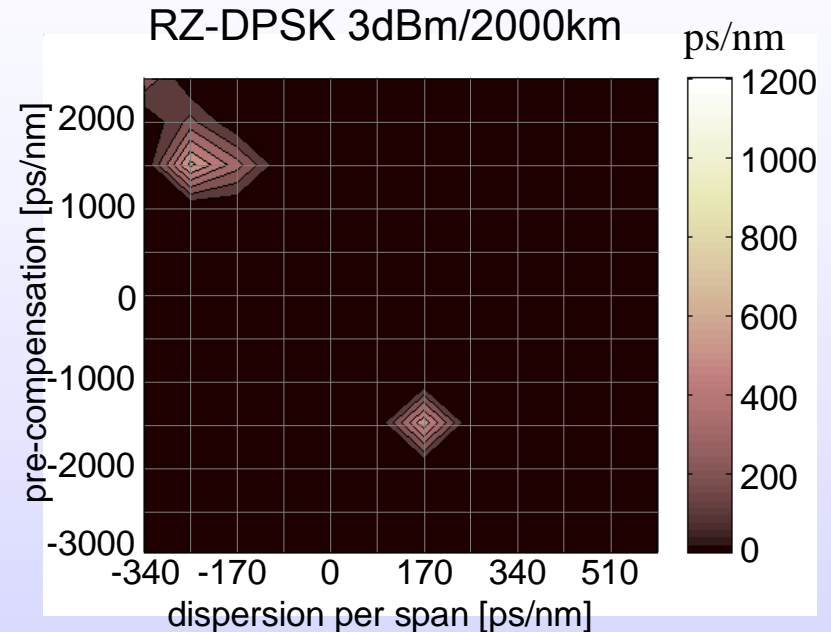
- One channel RZ-DQPSK or RZ-DPSK
- All other channels NRZ-ASK

Results for RZ-DPSK

uniform WDM-System:



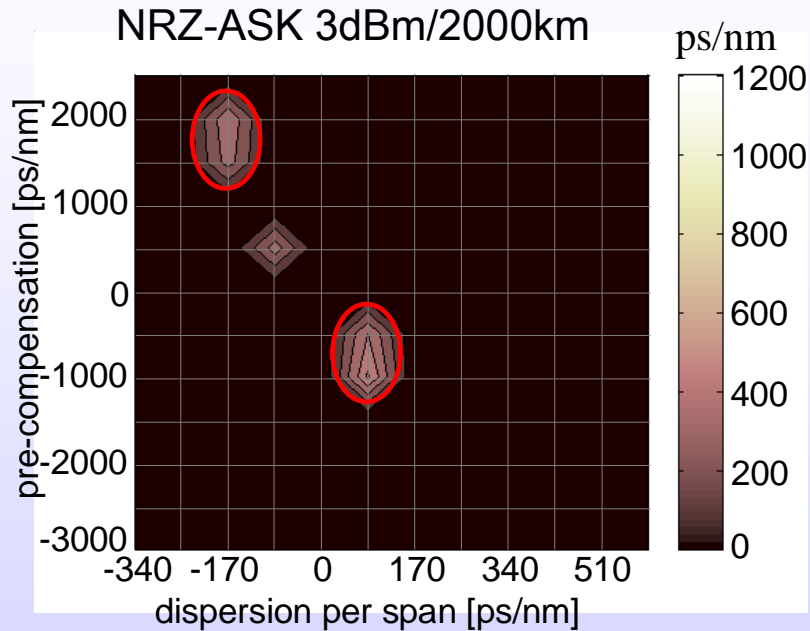
mixed WDM-System:



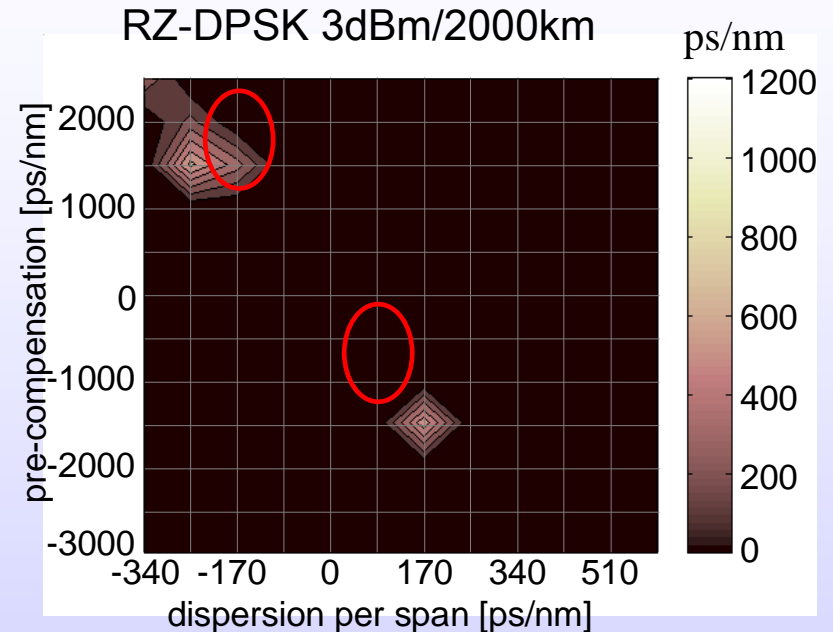
- Mixed WDM system: RZ-DPSK exhibits reduced dispersion tolerance

Results for RZ-DPSK

Uniform “old” WDM-System:



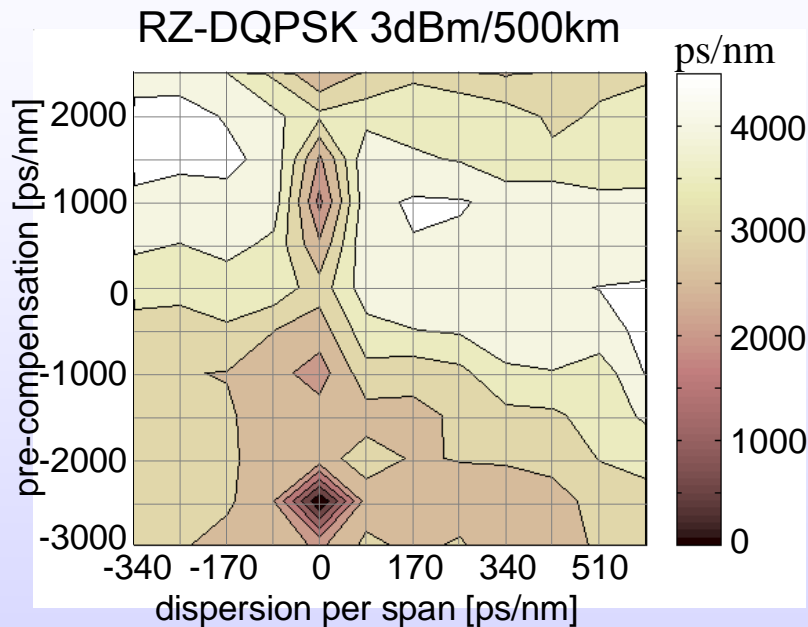
Mixed “upgraded” WDM-System:



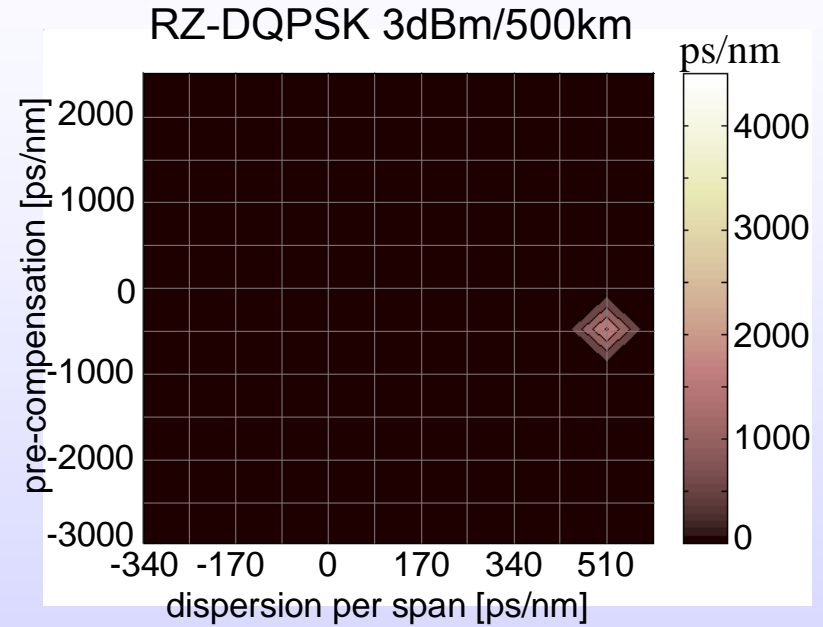
- Optimized Dispersion map for NRZ-ASK is sub-optimum for RZ-DPSK

Results for RZ-DQPSK

uniform WDM-System:



mixed WDM-System:

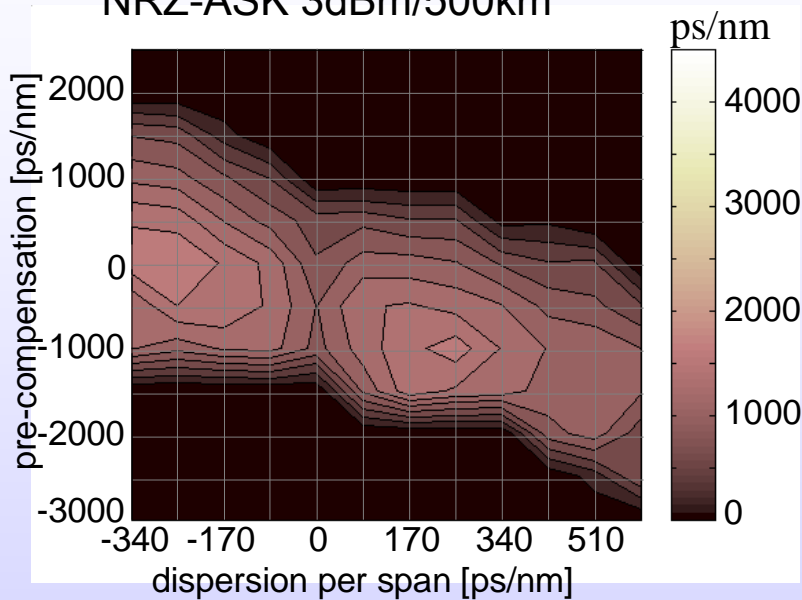


- Mixed WDM system: RZ-DQPSK exhibits strongly reduced dispersion tolerance

Results for RZ-DQPSK

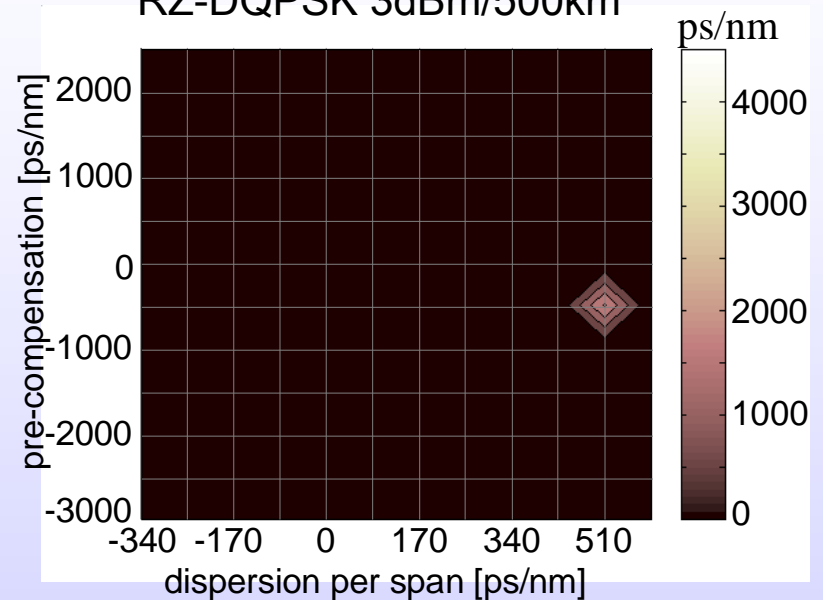
uniform WDM-System:

NRZ-ASK 3dBm/500km



mixed WDM-System:

RZ-DQPSK 3dBm/500km



- Optimized Dispersion map for NRZ-ASK is not suitable for RZ-DQPSK

Conclusions: Upgrading DWDM

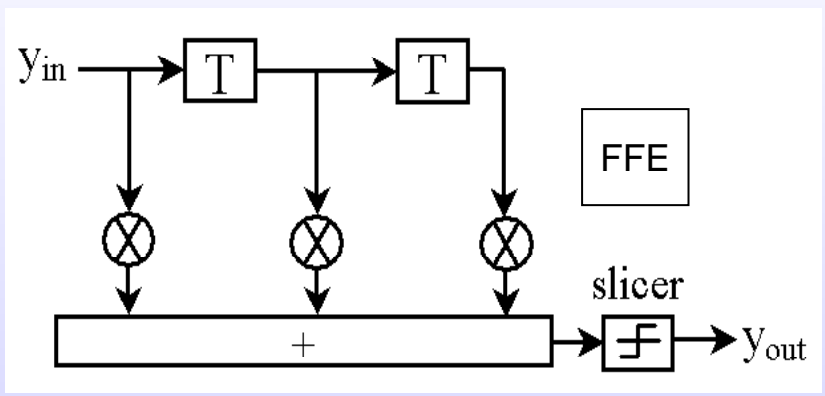
- In NRZ-ASK mixed modulated WDM-Systems strong degradation of RZ-DPSK and RZ-DQPSK
- RZ-DQPSK is degraded more than RZ-DPSK
- Transmission of RZ-DPSK and RZ-DQPSK in a mixed modulated NRZ-ASK WDM-System requires adaptation of Dispersion scheme

Electronic Equalizers (EDC) & Modulation Formats

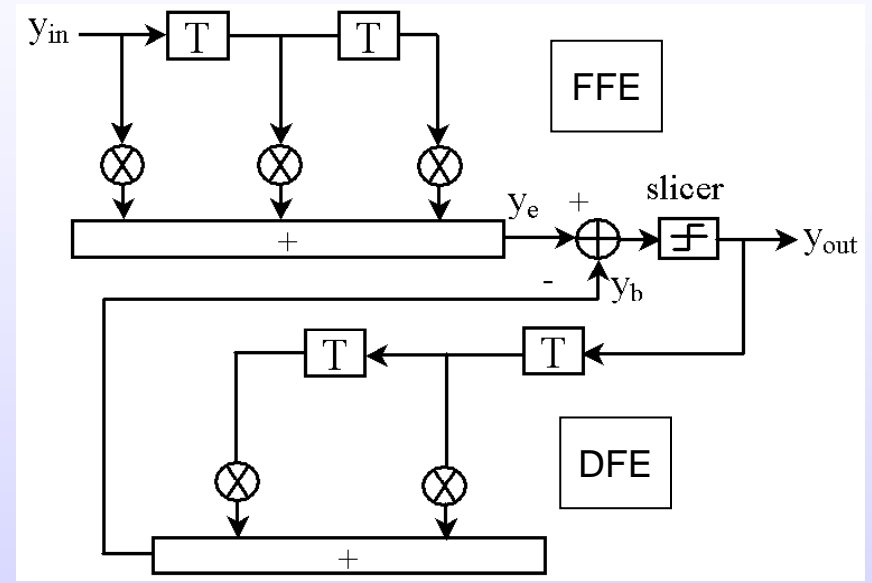
- Introduction of different EDC setups
- EDC on CD and PMD mitigation for advanced modulation formats
 - Optical duobinary modulation (ODB)
 - Differential-phase-shift-keyed (DPSK)
 - Optical single side band modulation (OSSB)
- Conclusions

EDC setups (1)

FFE



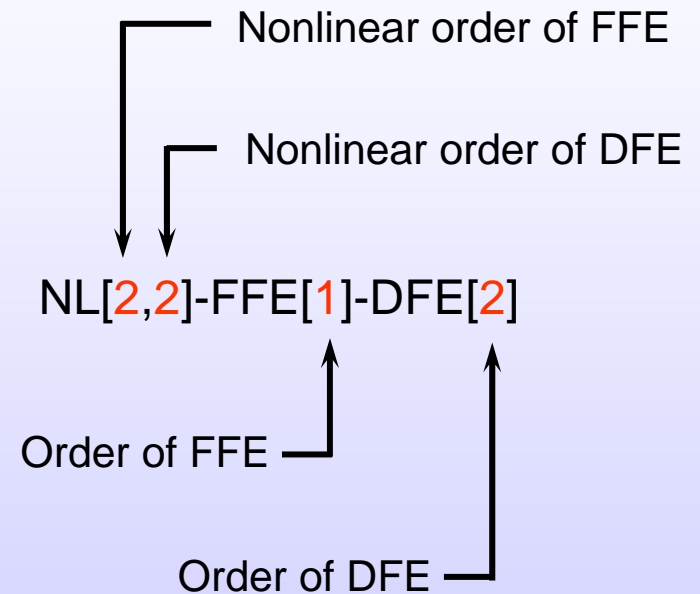
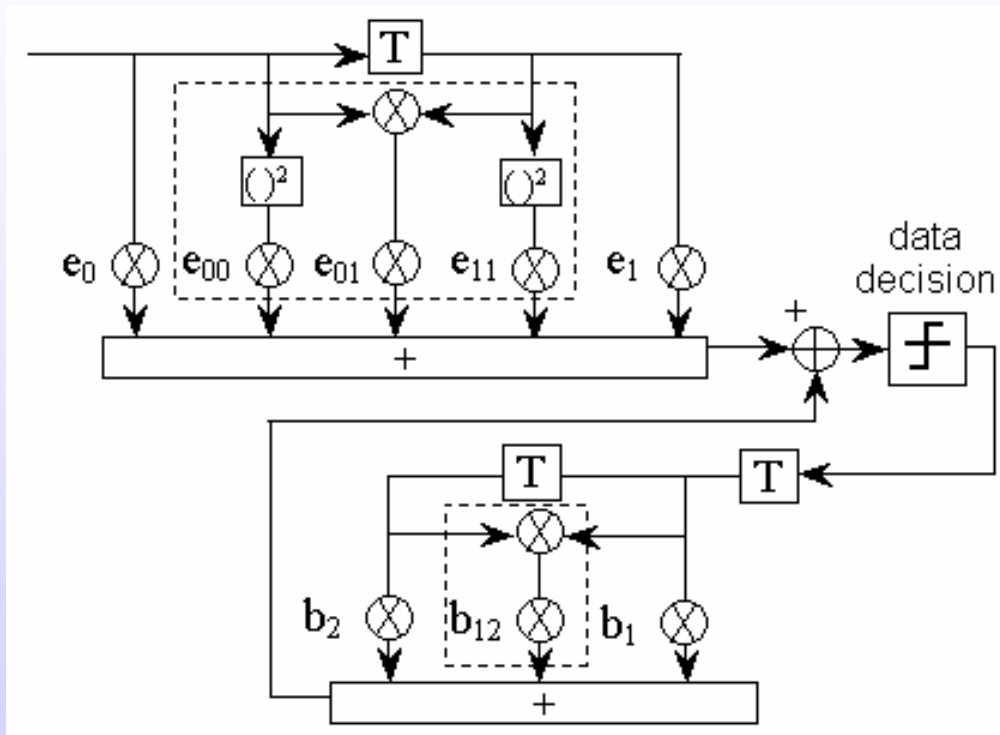
FFE-DFE



- Delay tap spacing: T (synchronous) or $T/2$ (fractionally spaced)
- Equalization coefficients are adaptively optimized, based on MMSE rule.

EDC setups (2)

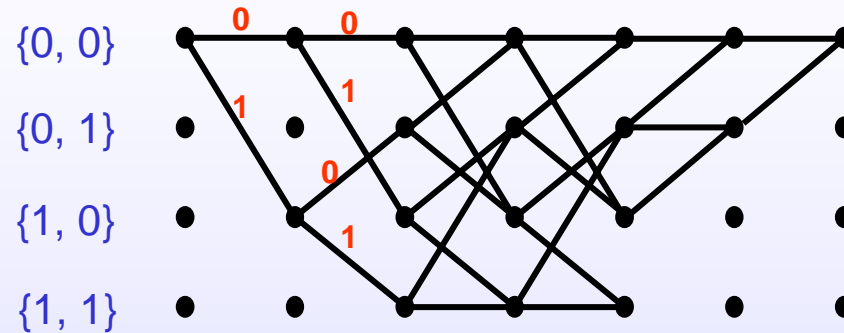
Nonlinear-FFE-DFE



- Extended from FFE-DFE including nonlinear ISI mitigation.
- Based on Voterra theory.

EDC setups (3)

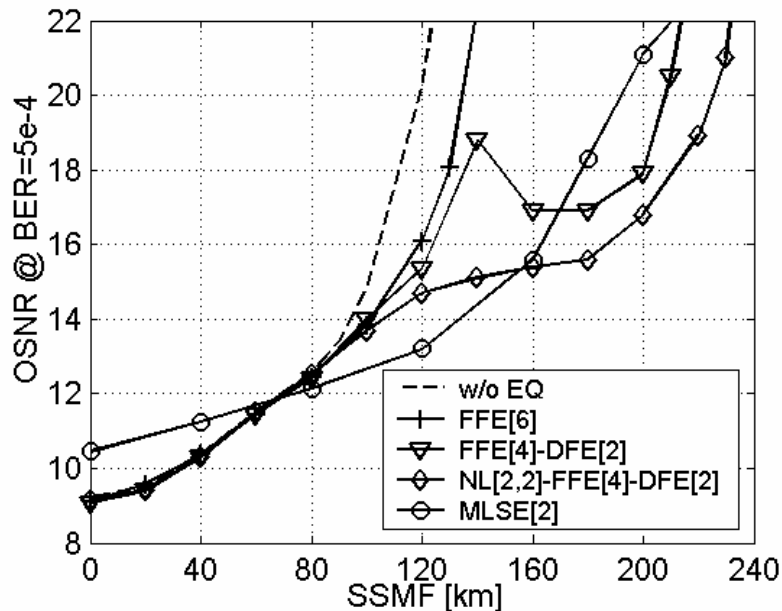
MLSE



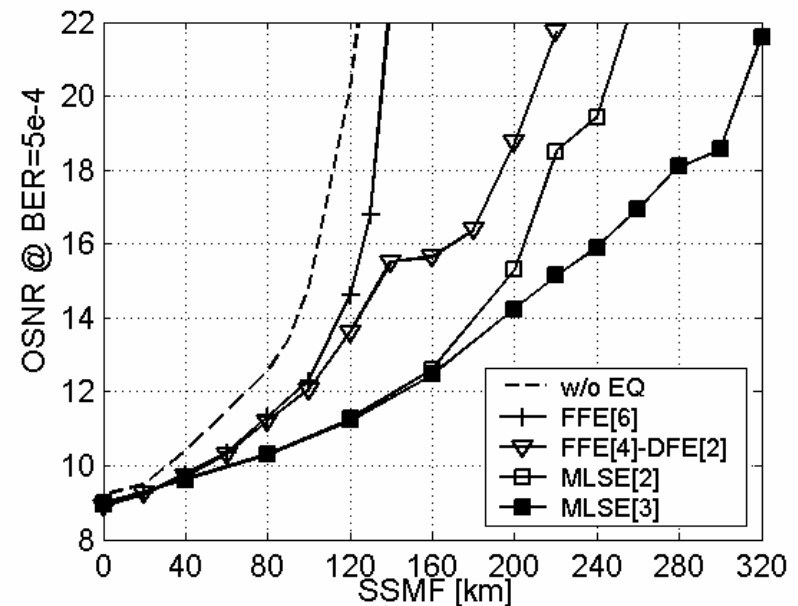
- Optimum receiver based on Viterbi algorithm
- Channel estimation (based on Lookup table method where PDF is estimated) first with training symbols.
- High speed A/D Converter required
- Memory: 2 or 3 (ISI from 2 or 3 previous bits)

EDC on CD mitigation for OOK

1 sample/bit, 10 Gb/s



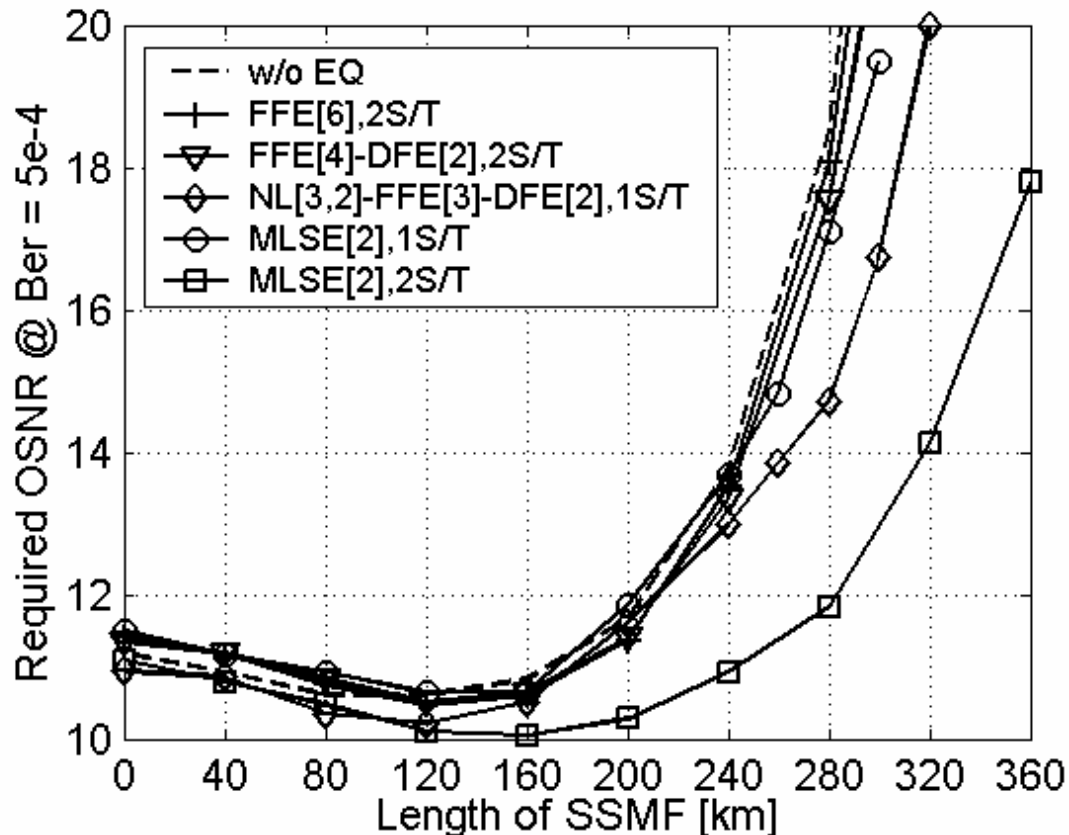
2 samples/bit, 10Gb/s



- FFE[4]-DFE[2] exhibits advantage over FFE[6] at longer distance
- NL[2,2]-FFE[4]-DFE[2] outperforms MLSE[2] at longer distance (>160km)

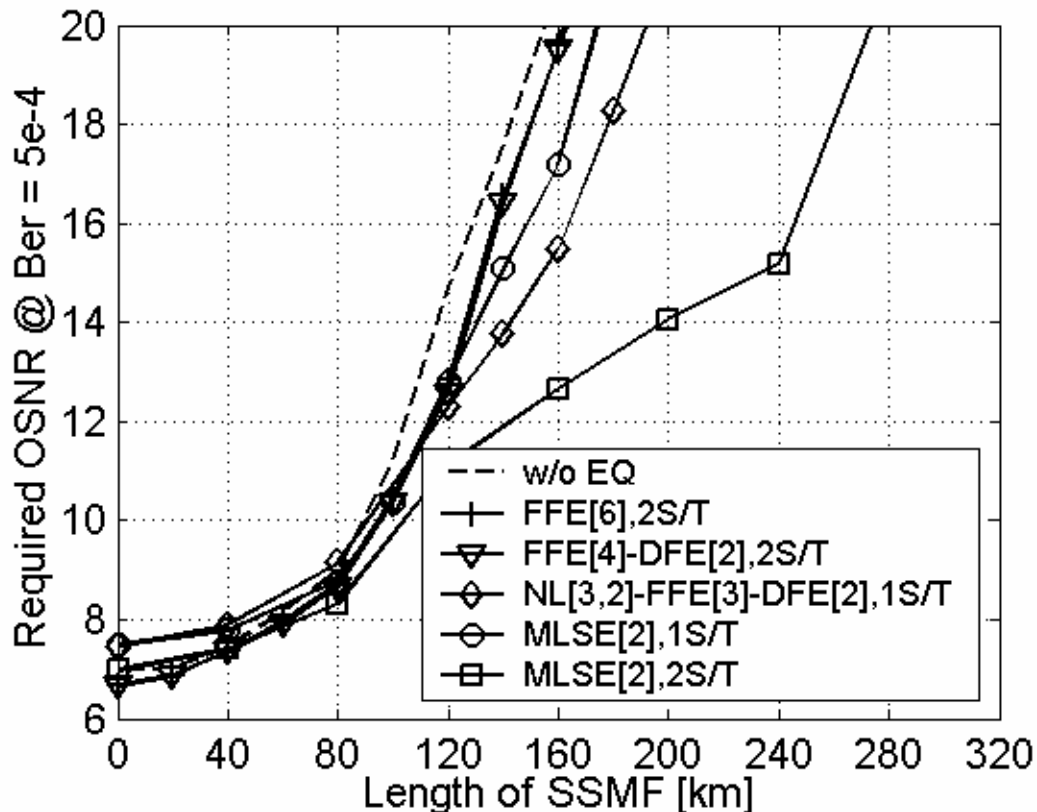
- Improvement through oversampling (fractionally spaced equalizer)

EDC on CD mitigation for **Optical Duobinary**



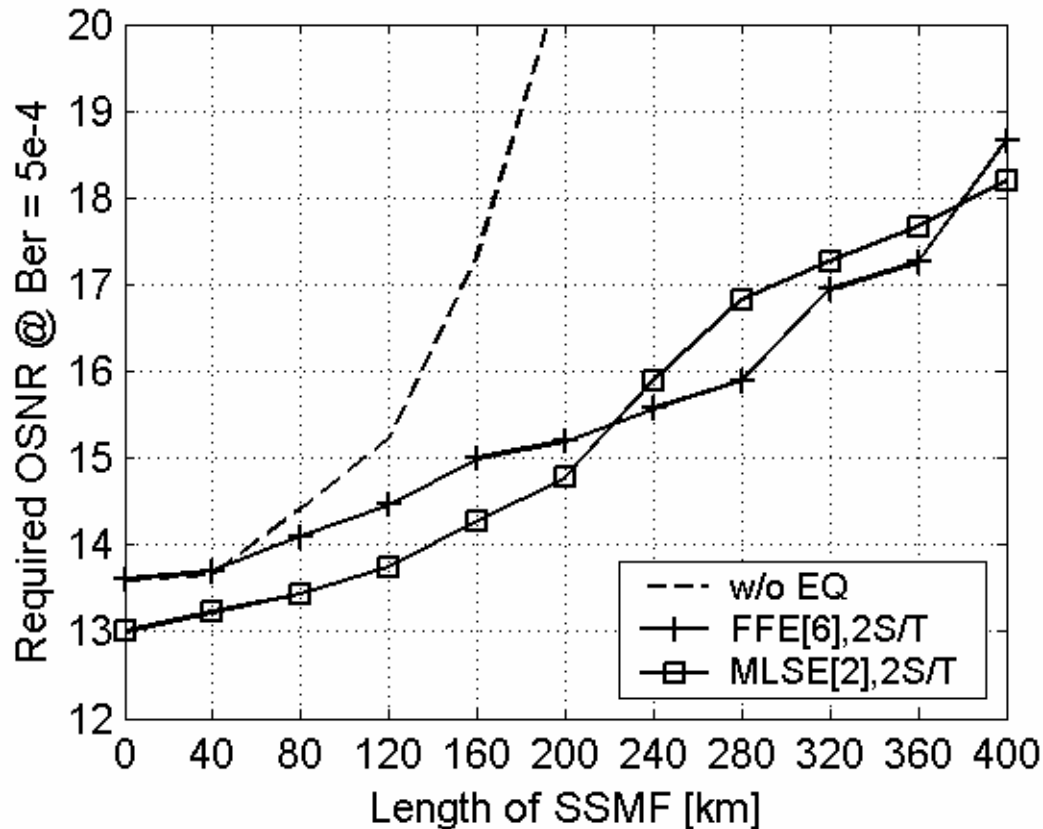
- ODB: larger dispersion tolerance, B2B penalty
- FFE and FFE-DFE very little performance improvement
- NL-FFE-DFE (1S/T) can achieve sub-optimum performance
- MLSE[2] with 2 sample/T required for better performance

EDC on CD mitigation for NRZ-DPSK



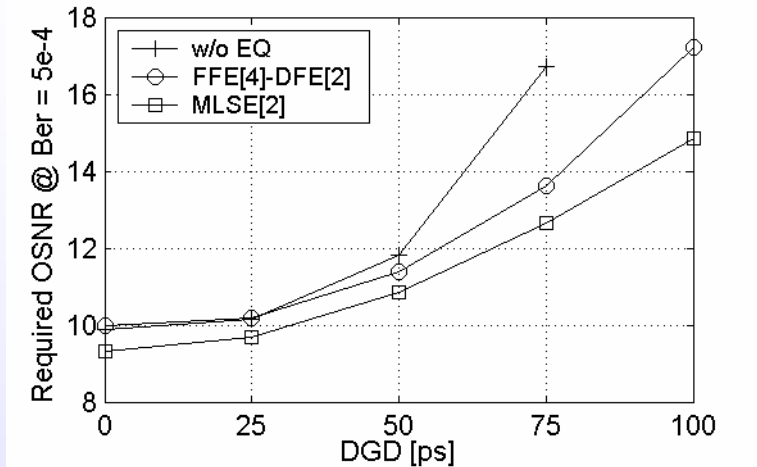
- NL-FFE-DFE outperforms MLSE[2] with 1sample/Bit
- MLSE[2] with 2samples/bit required for better performance
- All the EDC are limited at short distance due to the balanced detection

EDC on CD mitigation for OSSB

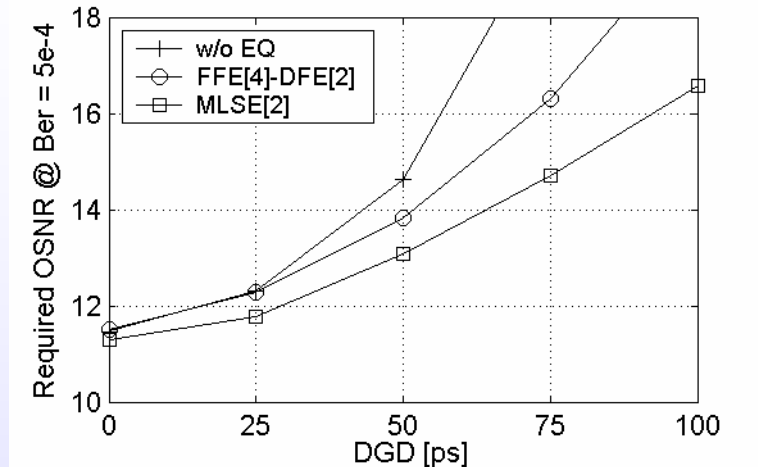


- CD results in Linear distortion in OSSB systems
- FFE[6] can achieve the similar performance to MLSE[2] up to 400km
- FFE alone can achieve very good compensation, DFE & MLSE are unnecessary

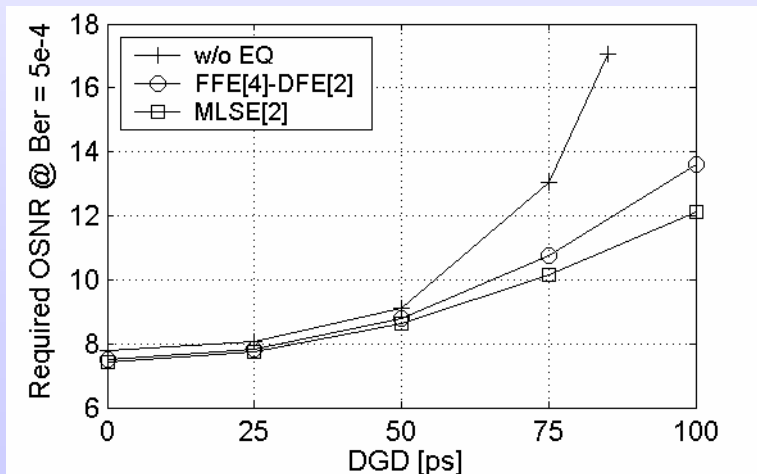
EDC on PMD mitigation for different modulation formats



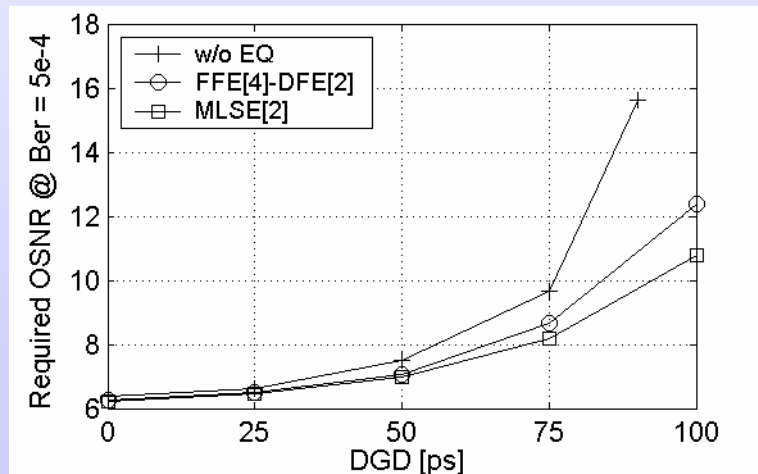
NRZ-OOK



NRZ-Duobinary



NRZ-DPSK



RZ-DPSK

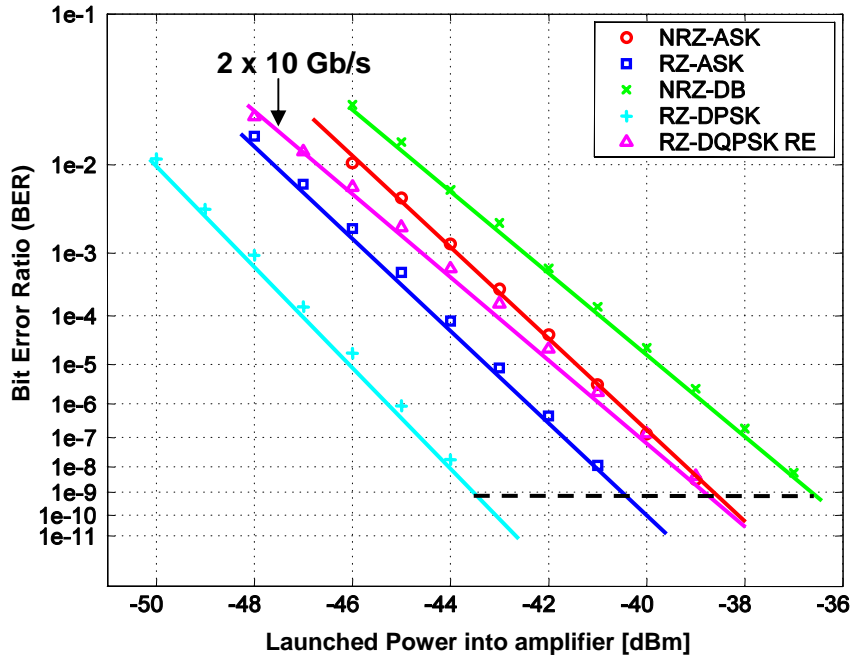
Conclusions: EDC on CD/PMD mitigation in SMF links

- EDC performance reduced for ODB and DPSK
- EDC with 2-fold oversampling better performance
- Nonlinear FFE-DFE shows nearly as good performance as MLSE.
- Good performance of EDC on PMD mitigation for all modulation formats

FEC & Modulation Formats

- Receiver Penalties and Net Coding Gain (NCG) for different Modulation Formats and FEC
- RZ-DQPSK + Concatenated FEC Codes

Receiver Penalty of Various Modulation Formats



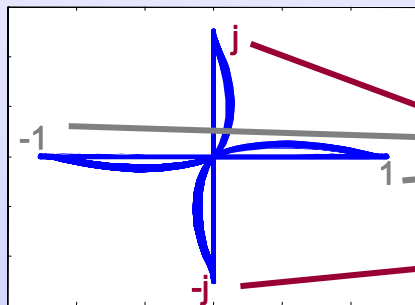
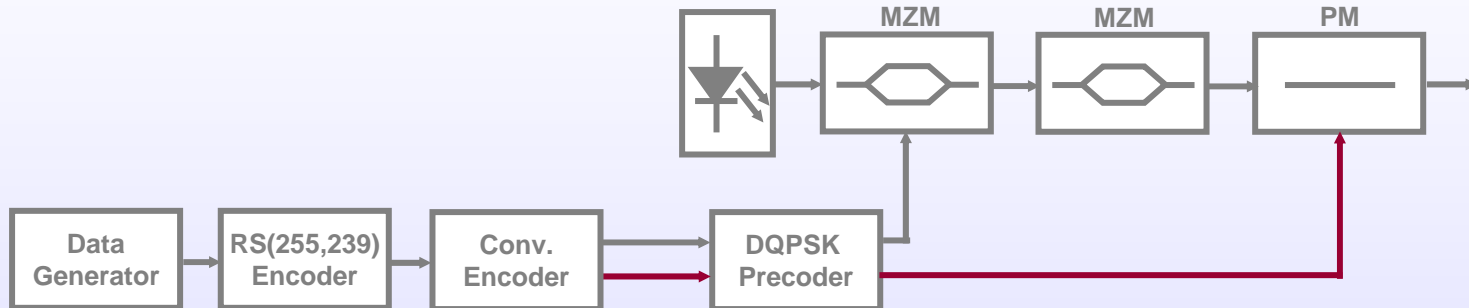
| Modulation Formats | Receiver Penalty without FEC at BER=10 ⁻⁹ | Receiver Penalty with FEC RS(255,239) NCG~4.8 dB | Receiver Penalty with FEC BCH(255,239) NCG~3.4 dB |
|--------------------|--|--|---|
| NRZ ASK | 0 dB (reference) | - 4.8 dB | - 3.4 dB |
| RZ ASK | - 1.8 dB | - 6.6 dB | - 5.2 dB |
| NRZ DB | + 2 dB | - 2.8 dB | - 1.4 dB |
| RZ DPSK | - 4.5 dB | - 9.3 dB | - 7.9 dB |
| RZ DQPSK | - 0.2dB | - 5 dB | - 3.6 dB |

- Monte Carlo Simulation, ASE noise
- NRZ ASK with FEC: 4.8dB NCG
- Modulation formats with different receiver sensitivity
- Coding Gain is added to all other Formats

RZ-DQPSK + Concatenated Codes

Standard RS(255,239) + R=1/2 Convolutional Code + RZ-DQPSK

- Serial concatenation of RS(255,239) (outer-code) and R=1/2 Convolutional (inner-code)



Signal Constellation of RZ-DQPSK

- In-phase and Quadrature Components

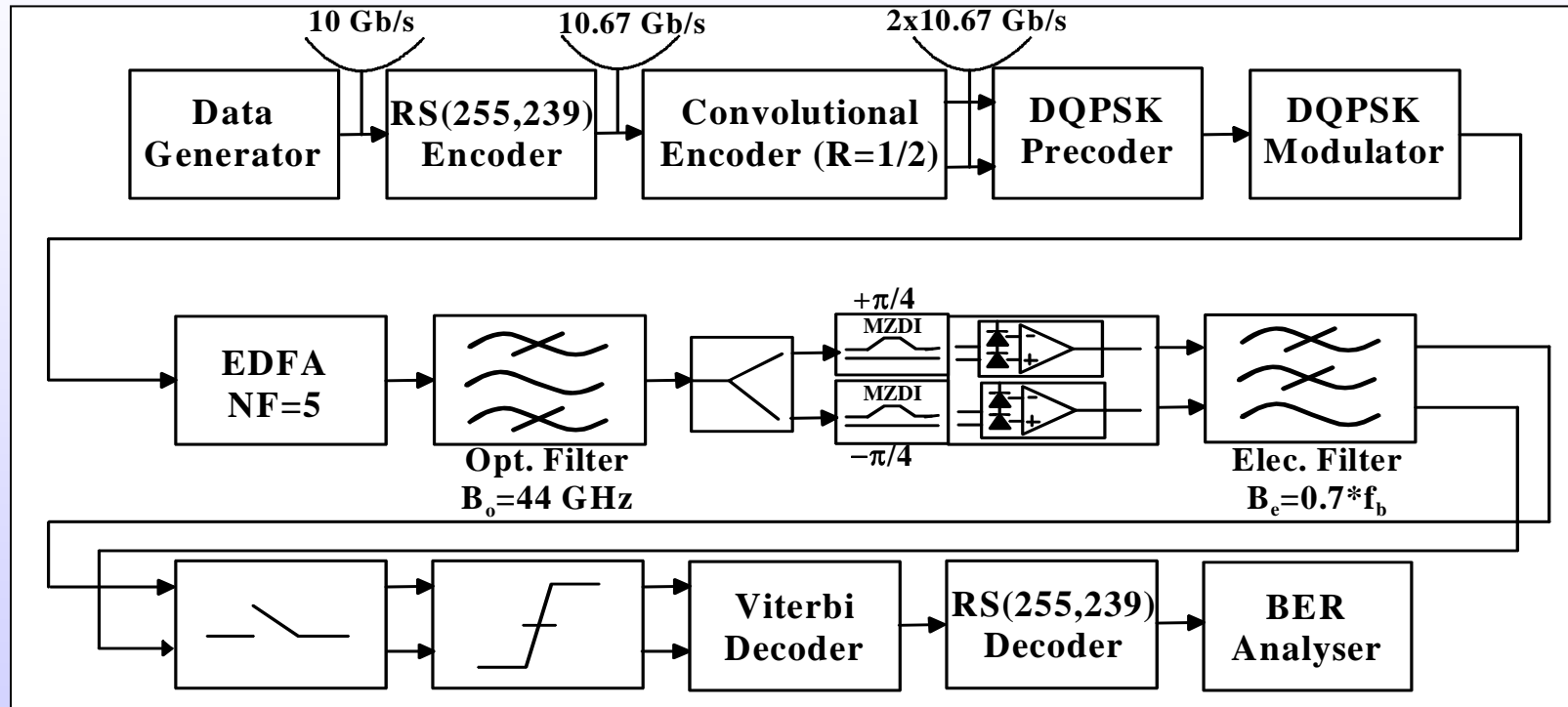
- Data Signal

- Overhead

- Data signal was allocated to the in-phase component
- Overhead was allocated to the quadrature component

Simulations System Setup

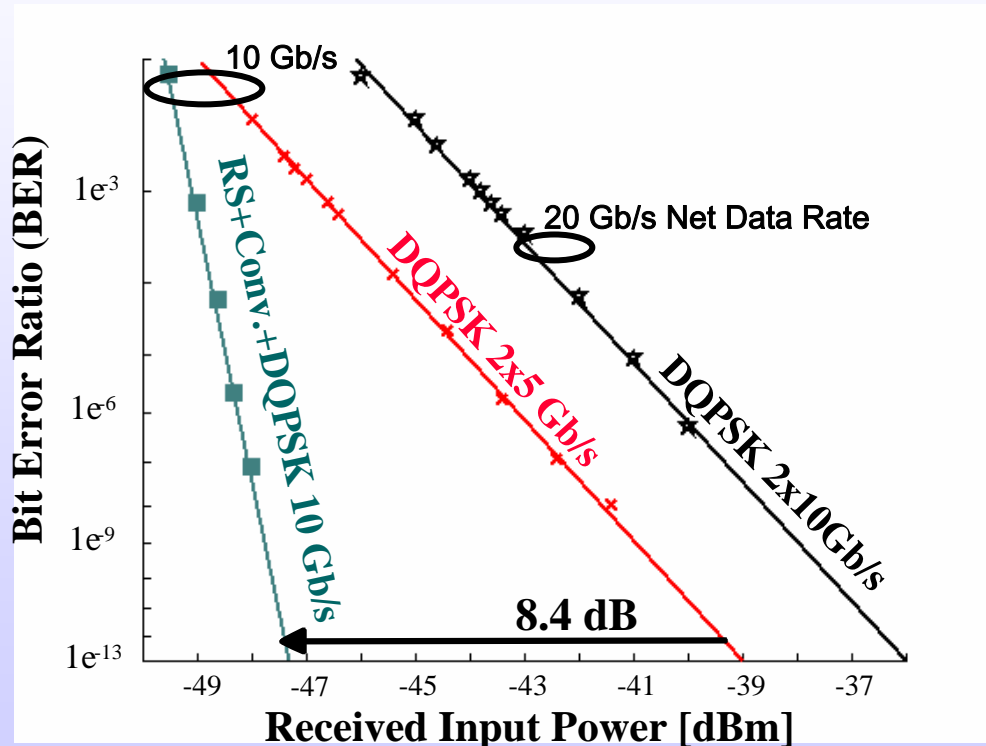
- Monte-Carlo simulations of a b2b optically preamplified system with dominating ASE noise
- RZ-DQPSK with serial concatenation of block and convolutional codes



- Interleaving was omitted in this concatenated scheme to simplify the simulations
- Hard-input hard-output Viterbi decoder was used

Bit Error Ratio Results

- Performance measured using BER-curves at 10^{-13}
- 6.7% OH \rightarrow Increase of bandwidth due to the standard RS(255,239) code
- 100% OH \rightarrow Increase of modulation levels (2 to 4) due to R=1/2 convolutional code



Results:

| | |
|---------------------------|---------------|
| <u>RS+RZ-DQPSK:</u> | 6.5 dB |
| <u>Conv.+RZ-DQPSK:</u> | 5.2 dB |
| <u>RS+Conv.+RZ-DQPSK:</u> | 8.4 dB |

8.4 dB Net Coding Gain with only 6.7% excess bandwidth over 2x5Gb/s DQPSK

Summary and Comments

- Comparison of several modulation formats
- Differences in Performance and Complexity
- Introducing advanced Modulation Formats in DWDM Networks:
 - Critical upgrade into existing IM/DD DWDM networks
 - Review equalizer design & performance
 - Enhanced capacity of multilevel modulation (as QPSK) may be used as overhead for error correction